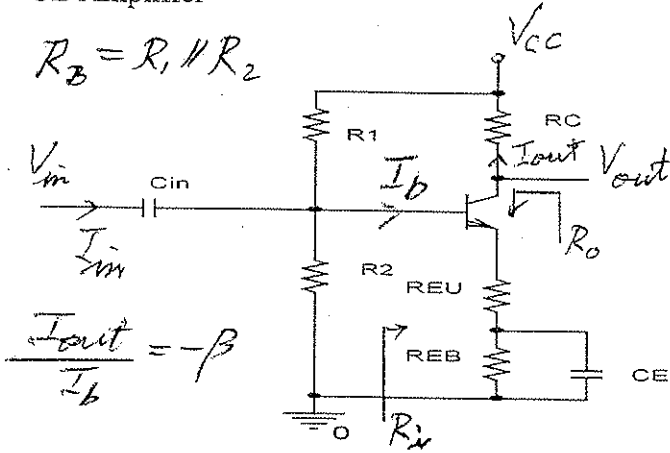


BJT Amplifier Configurations

CE Amplifier



Small-signal Analysis

$$A_I = \frac{I_{out}}{I_{in}} = -\beta \times \frac{R_B}{R_i + R_B}$$

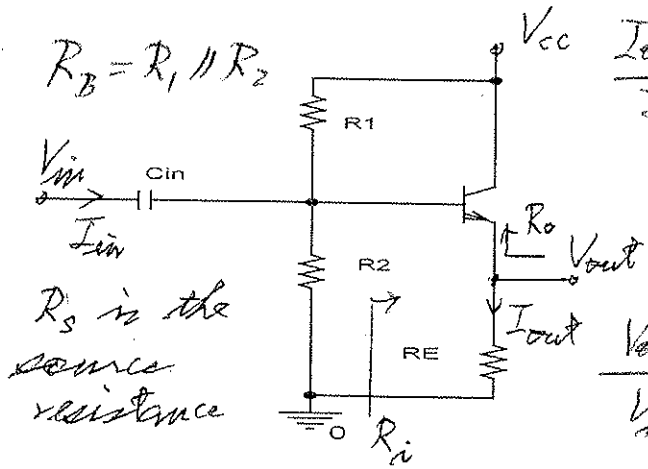
$$R_o = \infty$$

$$R_i = r_{\pi} + (1 + \beta) R_{EU}$$

$$A_V = \frac{V_{out}}{V_{in}} = \begin{cases} -g_m R_c & (R_{EU} = 0) \\ -\frac{\beta R_c}{\beta R_{EU} + R_{EU} + r_{\pi}} & (R_{EU} \neq 0) \end{cases}$$

$$A_V < 0$$

CC Amplifier (Emitter follower)



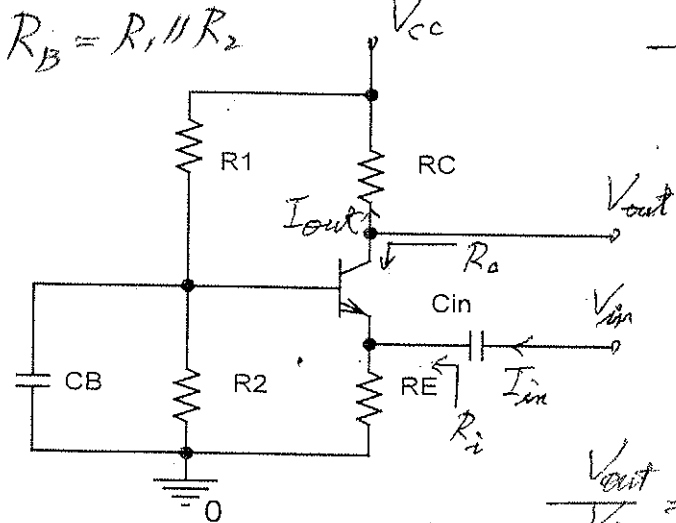
$$\frac{I_{out}}{I_{in}} = A_I = (\beta + 1) \times \frac{R_B}{R_B + R_i} \quad (R_s = 0)$$

$$R_o = \frac{(R_s \parallel R_B) + r_{\pi}}{1 + \beta} \approx \frac{1}{g_m} \quad (\beta \gg 1)$$

$$R_i = r_{\pi} + (1 + \beta) R_E$$

$$\frac{V_{out}}{V_{in}} = A_V = \frac{g_m + \frac{1}{r_{\pi}}}{\frac{1}{R_E} + g_m + \frac{1}{r_{\pi}}}$$

CB Amplifier



If $g_m + \frac{1}{r_{\pi}} \gg \frac{1}{R_E}$, then $A_V \approx 1$

with C_B

$$A_I = \frac{I_{out}}{I_{in}}$$

$$A_I = \frac{g_m}{\frac{1}{R_E} + \frac{1}{r_{\pi}} + g_m} \approx 1 \text{ if } g_m \gg \frac{1}{R_E} + \frac{1}{r_{\pi}}$$

$$R_o = \infty$$

$$R_i = \frac{1}{\frac{1}{R_E} + \frac{1}{r_{\pi}} + g_m} \approx \frac{1}{g_m} \text{ if } g_m \gg \frac{1}{R_E} + \frac{1}{r_{\pi}}$$

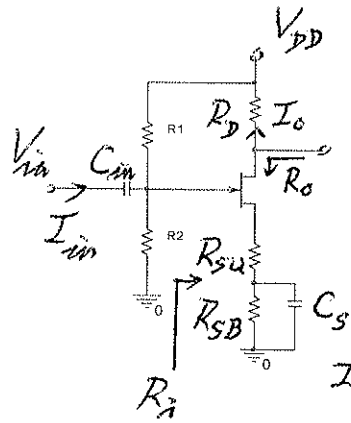
$$\frac{V_{out}}{V_{in}} = A_V = g_m R_c \quad (A_V > 0)$$

Without C_B $A_I = g_m \times \frac{r_{\pi}}{r_{\pi} + R_B} \times R_i$ where $R_i = \frac{1}{\frac{1}{R_E} + \frac{1}{r_{\pi} + R_B} + \frac{g_m r_{\pi}}{r_{\pi} + R_B}}$

$$R_o = \infty, \quad A_V = g_m R_c \times \frac{r_{\pi}}{r_{\pi} + R_B}$$

JFET Amplifier Configurations

Common Source Amplifier



$$A_V = \frac{V_o}{V_{in}} = - \frac{\mu R_D}{Y_o + R_D + (1 + \mu) R_{Su}} < 0$$

$$A_I = \frac{I_o}{I_{in}} = - \frac{\mu R_G}{Y_o + R_D + (1 + \mu) R_{Su}} < 0$$

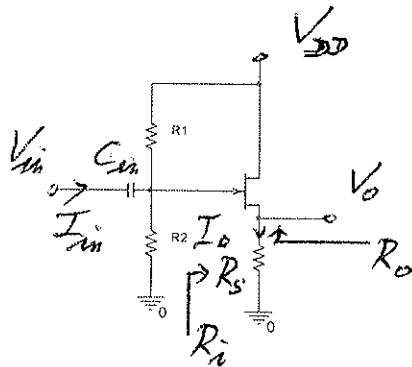
$$R_i = \infty, R_o = Y_o + (1 + \mu) R_{Su}$$

$$\text{If } Y_o = \infty, A_V = \frac{-g_m R_D}{1 + g_m R_{Su}}, R_o = \infty$$

$$R_G = R_1 || R_2, R_i = \infty, A_I = - \frac{g_m R_G}{1 + g_m R_{Su}}$$

$$\mu = g_m Y_o$$

Common Drain Amplifier (Source Follower)



$$A_V = \frac{V_o}{V_{in}} = \frac{g_m (Y_o || R_S)}{1 + g_m (Y_o || R_S)} > 0$$

$$A_I = \frac{I_o}{I_{in}} = A_V \times \frac{R_G}{R_S}$$

$$R_i = \infty, R_o = \frac{Y_o}{1 + g_m Y_o} = \frac{Y_o}{1 + \mu}$$

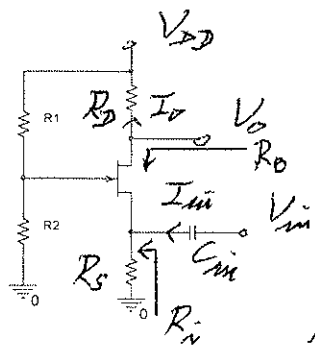
$$\text{If } Y_o = \infty, A_V = \frac{g_m R_S}{1 + g_m R_S}$$

$$A_I = \frac{g_m R_G}{1 + g_m R_S}, R_i = \infty$$

$$R_o = \frac{1}{g_m}$$

JFET Amplifier Configurations

Common Gate Amplifier



$$A_V = \frac{V_o}{V_{in}} = \frac{(1 + \mu) R_D}{R_D + Y_o}$$

$$A_I = \frac{I_o}{I_{in}} = \frac{A_V}{R_D \left(\frac{1}{R_S} + \frac{1}{Y_o} + g_m \right) - \frac{R_D}{Y_o} A_V}$$

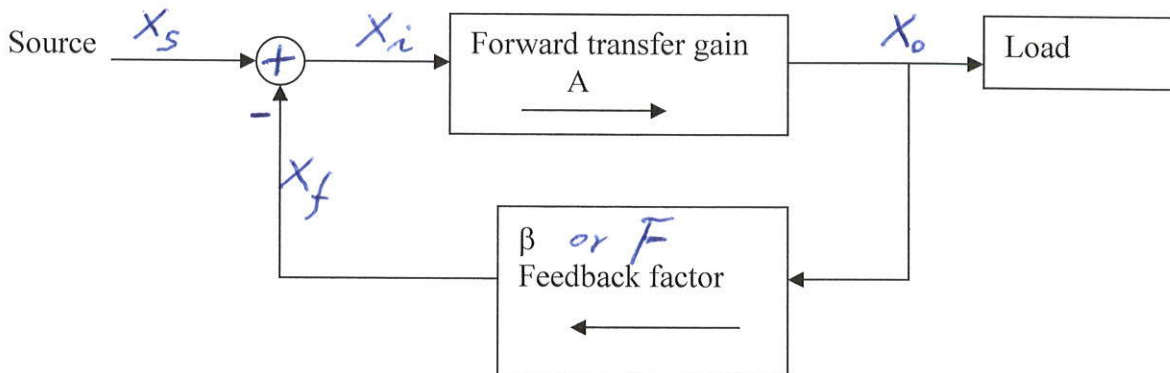
$$R_i = \frac{R_S Y_o}{Y_o + (1 + \mu) R_S - A_V R_S}, \quad R_o = Y_o$$

If $Y_o = \infty$, $A_V = g_m R_D$

$$A_I = \frac{g_m R_S}{1 + g_m R_S}, \quad R_o = \infty$$

$$R_i = \frac{R_S}{1 + g_m R_S}$$

Method of analyzing a negative feedback amplifier



(i) Identify the (ac) feedback topology

Input mixing

Input is series (voltage) mixing if you can indicate (formulate) $V_i = V_s - V_f$ (KVL). $X_s = V_s$.

Input is shunt (current) mixing if you can indicate (formulate) $I_i = I_s - I_f$ (KCL). $X_s = I_s$.

Output sampling

If $V_{out} = 0$ (Set $R_L = 0$), X_f (feedback signal) = 0. It is a voltage sampling at the output, $X_o = V_{out}$.

If $I_{out} = 0$ (Set $R_L = \infty$), X_f (feedback signal) = 0. It is a current sampling at the output, $X_o = I_{out}$.

Feedback factor = $F = X_f / X_o$. **Note that F should be independent of the load resistance.**

(ii) Obtain the ac equivalent circuit (A circuit) of the amplifier without feedback

1. Let dc supply voltage = 0 and connect all ground points together.
2. Delete the ground notation.

3. Remove the feedback signal from the output and redraw the input circuit.

Set $V_{\text{out}} = 0$ if it is a voltage sampling at the output (Short circuit the output node.).

Set $I_{\text{out}} = 0$ if it is a current sampling at the output (Open circuit the output loop.).

4. Remove the feedback signal to the input and redraw the output circuit. Indicate X_f on the output circuit (NOT the input circuit.)

Set $V_i = 0$ for current mixing.

Set $I_i = 0$ for voltage mixing.

(iii) Optional: Use a Thevenin's (Norton's) equivalent for the source if X_f is a voltage (current) variable.

(iv) Replace each active device by its small-signal model.

(v) Evaluate A , F (or β), input resistance and output resistance. $A = X_o / X_s$.

(vi) From A and F, determine the closed-loop (feedback) gain, input and output resistances of the original negative feedback amplifier.

$$A_f = \frac{A}{1 + FA}, \quad D = 1 + FA, \quad R_{i,f} = D R_i \text{ (voltage mixing) or } \frac{R_i}{D} \text{ (current mixing)}$$

$$R_{o,f} = \lim_{\text{load resistance} \rightarrow \infty} \frac{R_o}{D} \quad \text{for voltage sampling}$$

$$R_{o,f} = \lim_{\text{load resistance} \rightarrow 0} DR_o \quad \text{for current sampling}$$